

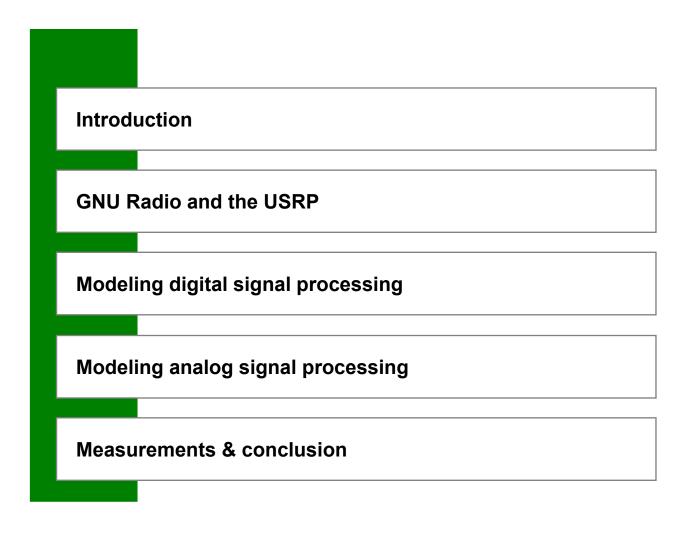
# Wireless Networks In-the-Loop: Emulating an RF front-end in GNU Radio

### Sebastian Koslowski, Martin Braun, Jens Elsner and Friedrich Jondral

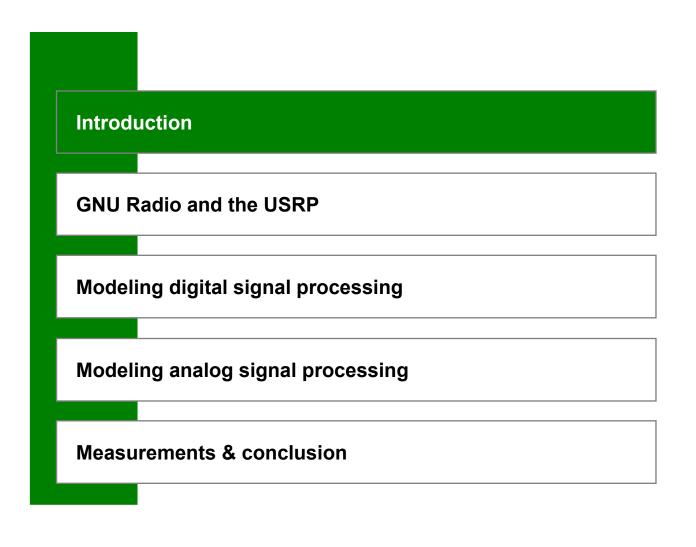
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# Introduction

Faster verification and development of SRs with loop simulation

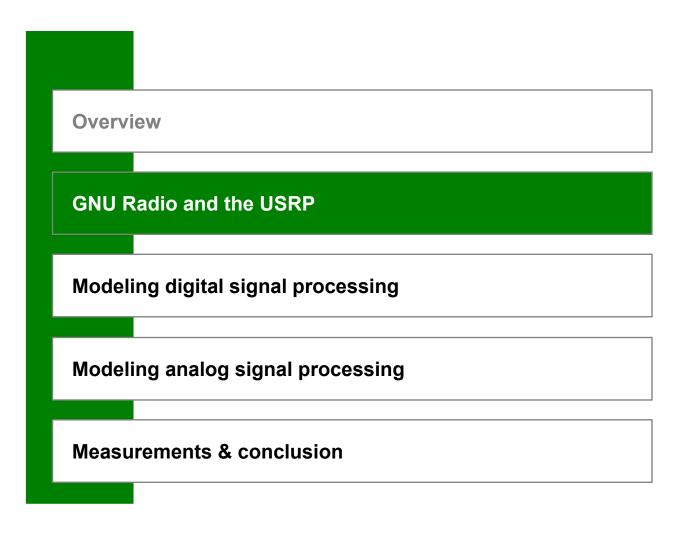
- For realistic results, effects of non-idealities need to be included in simulation
- Together with a channel emulator this facilitates loop simulation of software radio signal processing code
- Considered here: USRP with RFX2400 as an example of generic RF hardware





Emulator properties	Emulator Implementation
<ul> <li>Modular design</li> </ul>	
<ul> <li>Parametrizable</li> </ul>	<ul> <li>C++ / GNU Radio block</li> </ul>
<ul> <li>Discrete representation of analog signals at an equivalent rate of 128 MHz</li> </ul>	<ul> <li>USRP interface</li> </ul>







# **GNU Radio**

### Free Software Radio Framework

#### **Properties**

- Toolkit with extensive collection of atomic signal processing operations
- Signal sources and sinks offer interfaces to system and hardware
- USRP well supported

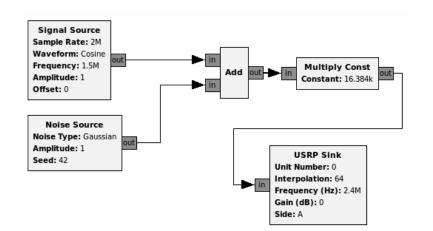
# GNU Radio

#### Design

- Signal processing in C++ (performance)
- Applications in Python (flow graphs)
- GNU Radio Companion offers graphical user interface

#### Execution

 Memory management and scheduling are handled by the GNU Radio framework



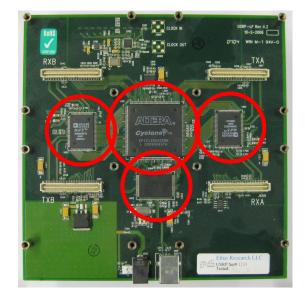
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# **Universal Software Radio Peripheral**

Ettus Research USRP1 as an example of generic RF hardware

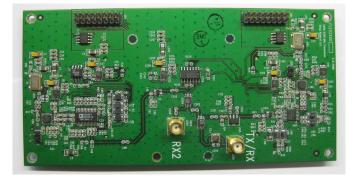
#### Motherboard

- Cypress FX2 USB Controller
   → Data exchange with host PC using USB 2.0
- Cyclone FPGA
   → Sample rate conversion and IF mixer
- Analog Devices AD9862 Mixed Signal Proc.
   → A/D conversion, interpolation, digital IF mixer

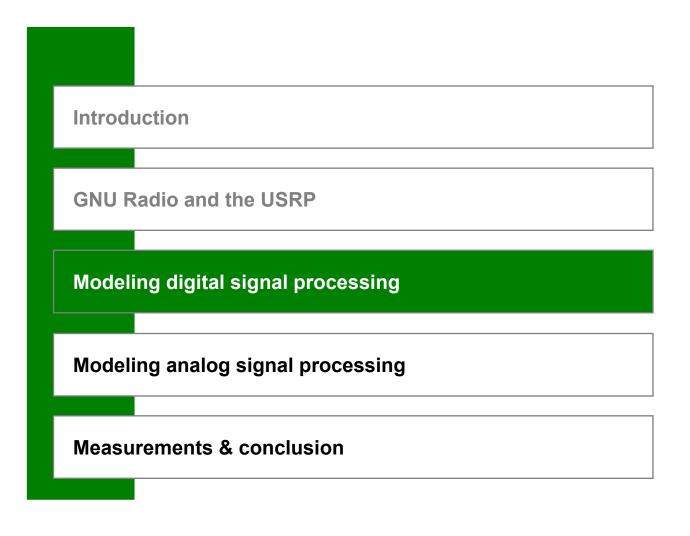


#### **RFX2400 Daughterboard**

- Direct conversion transceiver for 2.4 GHz band
- Two local oscillator signals, for Rx / Tx
- Maximum output power +17 dBm
- Integrated AGC (not modeled in emulator)



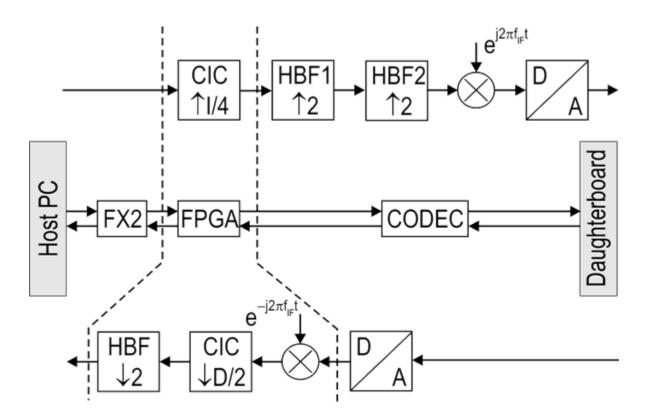






# Model of the USRP motherboards

#### Overview



#### Elements of digital signal processing

- A/D and D/A conversion
- Digital IF mixer
- Variable sample rate conversion with CIC- and halfband filters

# Model of the USRP motherboard

Processing in transmit and receive path

#### Analog/digital conversion

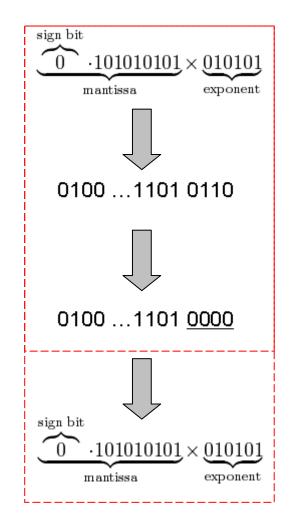
- Floating point to model analog signals in emulation
- A/D conversion with 12 bit/s and 64 MS/s
   → Decimation by factor 2 needed
- D/A conversion with 14 bit/s and 128 MS/s

 $\rightarrow$  Quantization noise / bit resolution by rounding

#### **Encoding of samples**

- FPGA and CODEC use fix point arithmetic
- GNU Radio framework offers high performance filters for floating point
- In emulator: floating point arithmetic for digital filters an mixers

 $\rightarrow$  Conversion to fix point between stages



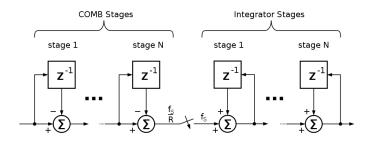
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# Model of the USRP motherboard

Processing in transmit and receive path

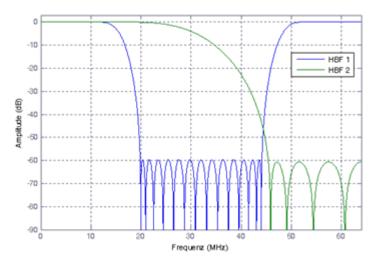
#### **Cascaded Integrator Comb filter**

- Efficient, recursive implementation of a multi-stage moving average filter for variable sample rate conversion
- USRP uses 4-stage CIC filter
- → Implemented in fix point arithmetic as GR block, scaling of output signal by bit shifting



#### Halfband filter

- Efficient FIR low pass filter; cut-off at band middle
- About half of all coefficients are zero
- $\rightarrow$  Used to interpolate / decimate by a factor of 2
- Tx: built as in data sheet
- Rx: coefficients as in USRP1 FPGA code

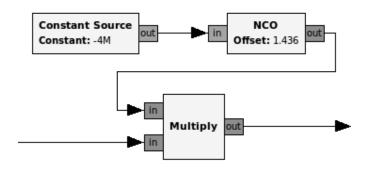


# Model of the USRP motherboard

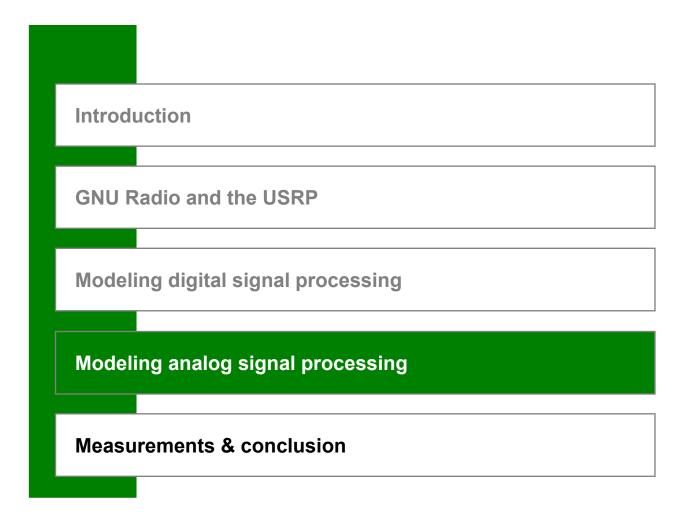
Processing in transmit and receive path

#### **Digital IF mixer**

- Digital mixer based on CORDIC algorithm
- Shifts base band signal to digital IF
- Together with analog mixer used to shift signal to carrier frequency
- → In emulator: complex multiplication with NCO, random initial phase

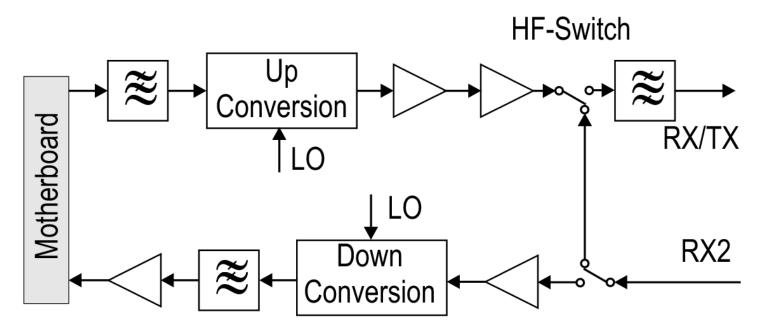








#### **Overview**



#### Typical effects of analog signal processing

Phase noise

Non-linearities

CEL

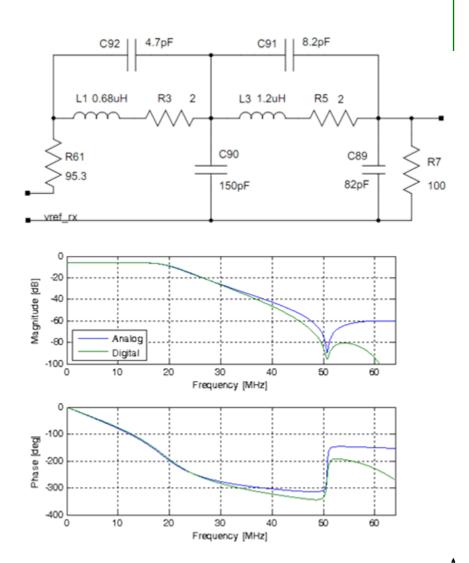
IQ-imbalance

Noise

### **Discretization of analog filters**

#### Anti aliasing filter

- After down shifting, before A/D conversion
- 5th order IIR low pass filter, cut-off frequency 20 MHz, attenuation 6 dB
- Discretization with bilinear transform
- Additional tweaking of poles and zeros due to aliasing effects needed
- → Approximated well in pass band, higher attenuation in stop band



CEL

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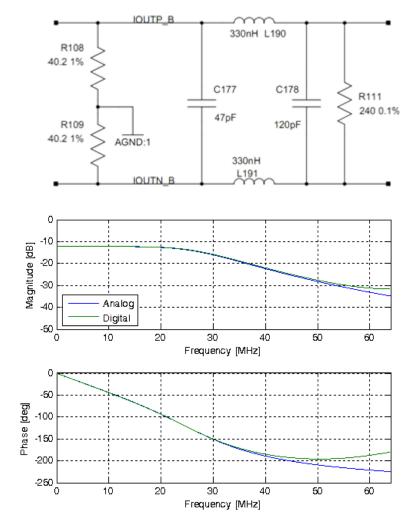
### **Discretization of analog filters**

#### Anti image filter

- After D/A conversion, before mixer
- 3rd order IIR low pass, cut-off frequency ~28 Mhz, attenuation 12 dB
- Discretization with impulse invariance method

#### **ISM bandpass filter**

- Magnitude response flat in simulation bandwidth
- Phase response not specified
- $\rightarrow$  Not modeled



CEL

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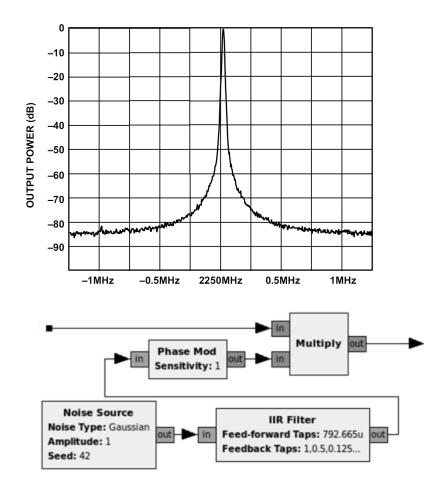
#### Phase noise

#### Phase noise

- LO signals of analog mixer are not ideal
   → 1/f-noise
- Created with noise shaping filter

$$H_{\mathsf{PN}}(z) = \frac{1}{\sqrt{1 - z^{-1}}}$$

- Approximation of IIR low pass coefficients with polynomial expansion
- In emulator: phase noise modulated onto the signal in baseband



### **IQ** imbalance

#### IQ imbalance

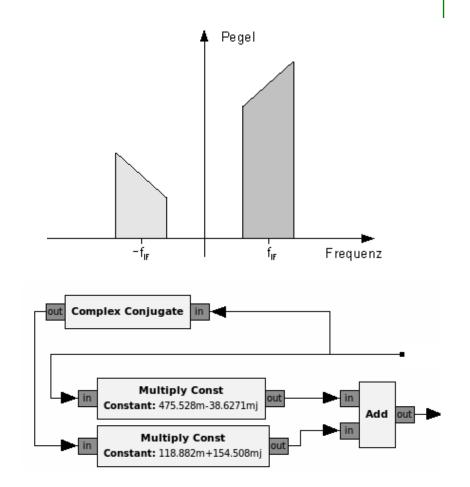
Phase shift in mixer is not perfect
 → Variations in amplitude and phase

$$r'(t) = (1 - \varepsilon)r_{bp}(t)\cos(\omega_c t + \phi) - j(1 + \varepsilon)r_{bp}(t)\sin(\omega_c t - \phi)$$

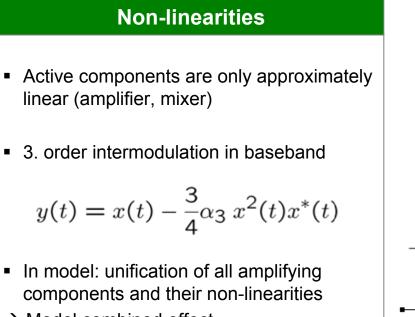
Image spectrum not suppressed fully

$$r'(t) = c_1 \cdot r(t) + c_2 \cdot r^*(t)$$

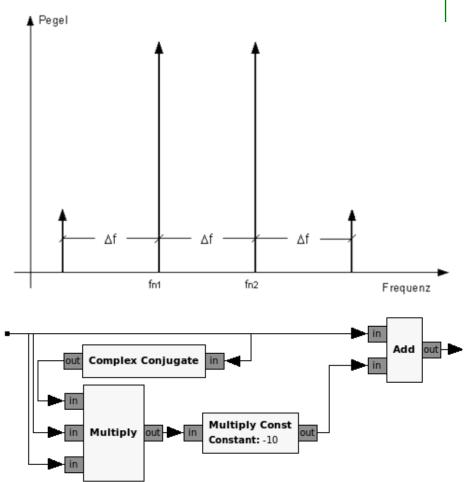
$$c_1 = \frac{1}{2}(\cos(\phi) - j\varepsilon\sin(\phi))$$
$$c_2 = \frac{1}{2}(\varepsilon\cos(\phi) + j\sin(\phi))$$



#### Intermodulation products



- → Model combined effect
- For baseband amplification quadratic mixer term is also relevant



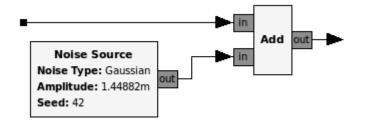
### Noise figure of receiver

#### **Receiver noise**

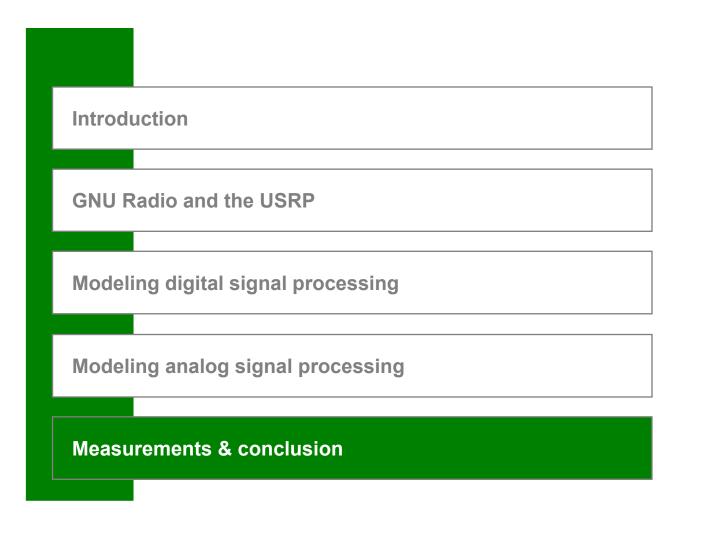
- Every component adds noise to signal and reduces SNR along the analog processing chain
  - $\rightarrow$  Characterization by noise figure
- Combining noise figures and gain of all components with Friis' formula
- In model: Johnson-noise is additive white Gaussian noise
- → NF  $\geq$  5dB at RX2 port
- NF depends strongly on AGC level

$$NF = 10 \log F = SNR_{in,dB} - SNR_{out,dB}$$

$$F_{\text{total}} = 1 + (F_1 - 1) + \frac{F_2 - 1}{g_1} + \frac{F_3 - 1}{g_1 g_2} + \dots$$



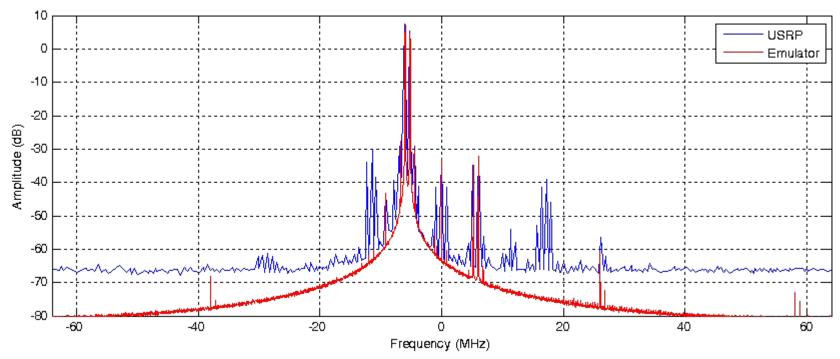






# **Measurements and comparison**

### Transmit path

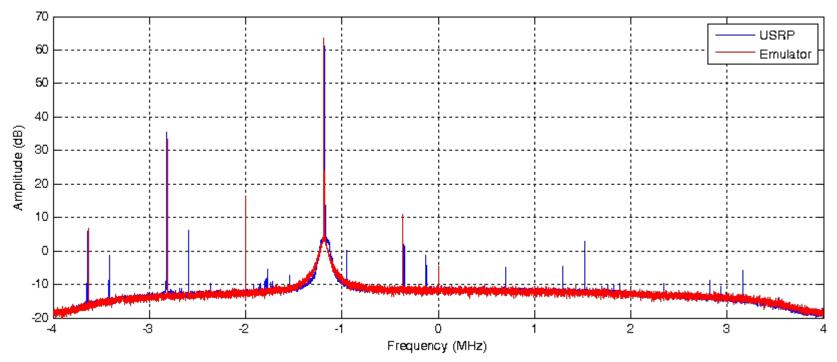


#### **Measurement setup**

- 2 tone signal with 800 kHz separation, carrier frequency 2.45 GHz, IF at -6 MHz
  - $\rightarrow$  IQ imbalance, carrier, IM products of 3rd order, phase noise, quantization noise visible
- Not included: thermal noise, additional non-linearities

### **Measurements and comparison**

### **Receive path**



#### **Measurement setup**

- Sinusoid at 1.2 MHz below carrier frequency, IF at -2 MHz.
  - $\rightarrow$  IQ imbalance, baseband non-linearities of 2nd order, phase noise



# Conclusion

- Modeled the core elements of the signal processing chain
  - Variable sampling rate conversion with CIC- and halfband filters
  - Digital IF mixing
  - D/A & A/D conversion (limited sample resolution, quantization noise)
  - Analog filters
- Modeled effects of analog RF hardware
  - Phase noise
  - IQ imbalance
  - Receiver noise
  - Non-linearities

# **Q&A / Discussion**

### Thank you for your attention!

